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Multi-physics simulation of mm-wave systems

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gan-on-Silicon Efficient mm-wave euRopean systEm iNtegration plAtform



Contributors

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 - I. Ndip, U. Maaß







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P Landin Dr. K. Anders





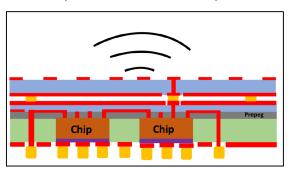




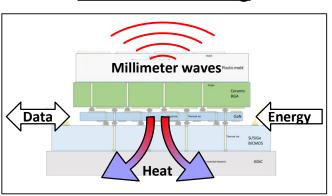
mm-wave RF systems

- Active antenna arrays
- High integration needed to fit within $\lambda/2$

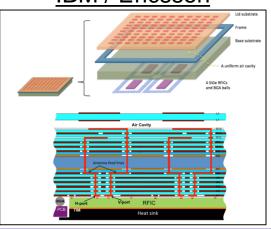
Chip embedding (IZM/SERENA)



Vertical stacking



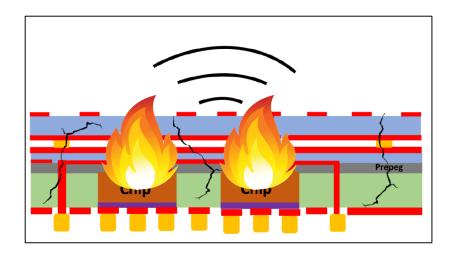
IBM / Ericsson





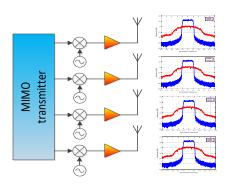
Challenges

- Multi-physical effects
 - Electrical
 - Thermal
 - Mechanical
 - **•** ...



- Efficient simulations and modeling are crucial
 - Before: Optimization, reliability, margins, time to market
 - After: Troubleshooting, performance optimization, reverse engineering

ELECTRICAL SIMULATIONS



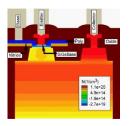
Active antenna arrays
Antenna-circuit interactions
Linearity



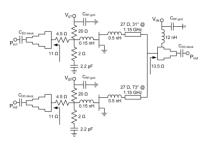


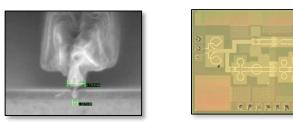
Simulations at many levels...

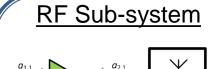
Transistor

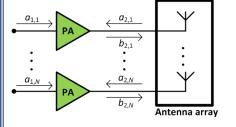


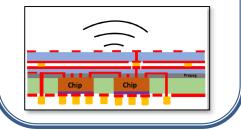
Circuit











<u>System</u>

$$y = Hx + w$$

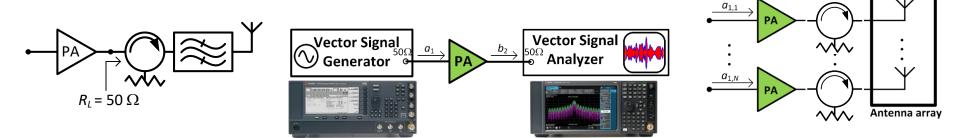


[T. Svensson, Chalmers]

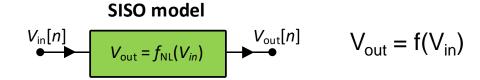


Transmitter modeling

• Traditionally 50Ω assumed



Single-input-single-output modeling of RF components



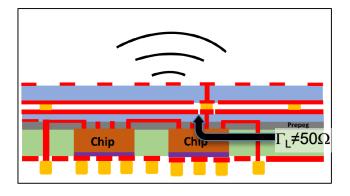


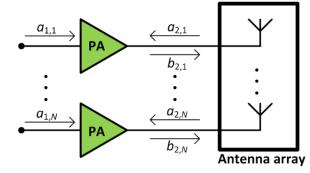


Transmitter modeling

- Integrated transmitters
 - Mismatch and mutual coupling
 - Non-50 Ω interfaces

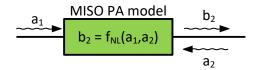
• Dual-input models needed: $b_2 = f(a_1, a_2)$







Dual-input behavioral modeling



$$b_{2} = \sum_{p_{1}=1}^{P_{1}} \alpha_{p_{1}} |a_{1}|^{2(p_{1}-1)} a_{1} + \sum_{p_{2}=1}^{P_{2}} \beta_{p_{2}} |a_{1}|^{2(p_{2}-1)} a_{2} + \sum_{p_{2}=2}^{P_{2}} \gamma_{p_{2}} a_{1}^{2} |a_{1}|^{2(p_{2}-2)} a_{2}^{*}$$

$$S_{21}(|a_{1}|)$$

$$S_{22}(|a_{1}|)$$

Nonlinear 50Ω SISO-terms

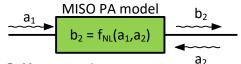
Nonlinear mismatch terms

"PHD" or "X-parameter®" model

[Verspecht and D. E. Root, "Polyharmonic distortion modeling," IEEE Microw. Mag., vol. 7, no. 3, pp. 44-57, Jun. 2006.]



Dual-input behavioral modeling



"X-parameters®" with memory

$$b_{2}[n] = \sum_{m_{1}=0}^{M_{1}} \sum_{p_{1}=1}^{P_{1}} \alpha_{m_{1},p_{1}} |a_{1}[n-m_{1}]|^{2(p_{1}-1)} a_{1}[n-m_{1}] + \sum_{m_{2}=0}^{M_{2}} \sum_{m_{1}=0}^{M_{1}} \sum_{p_{2}=1}^{P_{2}} \beta_{m_{1},m_{2},p_{2}} |a_{1}[n-m_{1}]|^{2(p_{2}-1)} a_{2}[n-m_{2}] + \sum_{m_{2}=0}^{M_{2}} \sum_{m_{1}=0}^{M_{1}} \sum_{p_{2}=2}^{P_{2}} \gamma_{m_{1},m_{2},p_{2}} a_{1}^{2}[n-m_{1}] |a_{1}[n-m_{1}]|^{2(p_{2}-2)} a_{2}^{*}[n-m_{2}]$$

$$T_{22}(a_{1})$$

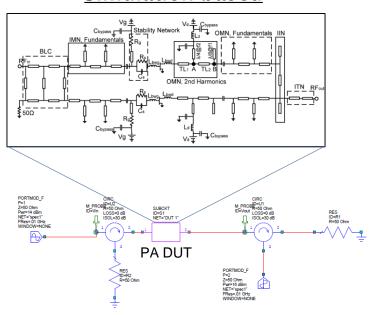
[C. Fager et al., "Prediction of Smart Antenna Transmitter Characteristics Using a New Behavioral Modeling Approach," Proc. IMS, 2014]



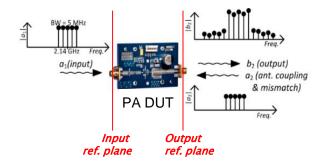


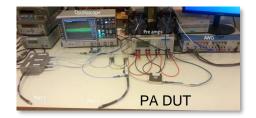
Model identification

Simulation based



Measurement based





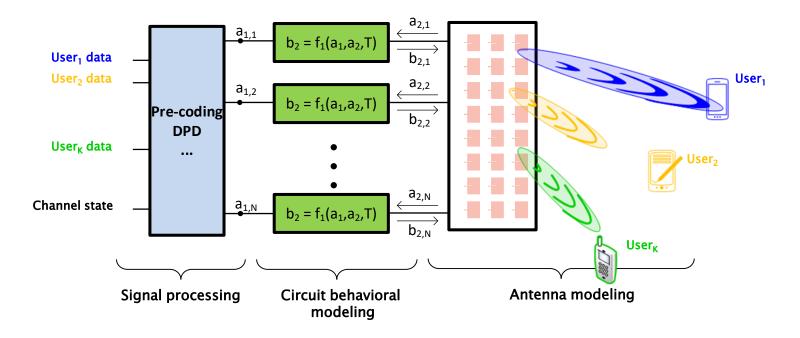
[S. Gustafsson et al., "A Novel Active Load-pull System with Multi-Band Capabilities," ARFTG, 2013]

[C. Fager et al., "Prediction of Smart Antenna Transmitter Characteristics Using a New Behavioral Modeling Approach," Proc. IMS, 2014]





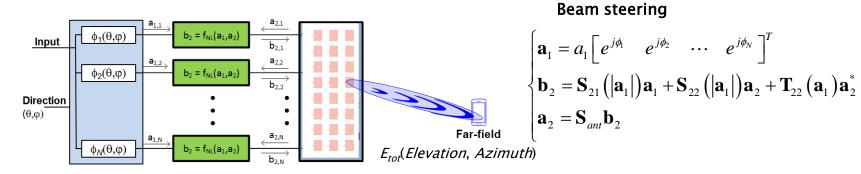
Transmitter simulation framework







Phased array application



Antenna circuit interactions

$$\mathbf{b}_{2} = \mathbf{S}_{21} (|\mathbf{a}_{1}|) \mathbf{a}_{1} + \mathbf{S}_{22} (|\mathbf{a}_{1}|) \mathbf{S}_{ant} \mathbf{b}_{2} + \mathbf{T}_{22} (\mathbf{a}_{1}) \mathbf{S}_{ant}^{*} \mathbf{b}_{2}^{*}$$
Regular 50\(\Omega\) Antenna coupling and mismatch effects

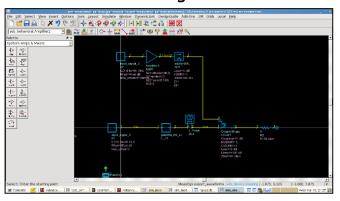
Far field radiation

$$E_{tot}(El,Az)[n] = \sum_{i=1}^{N} b_{2,i}[n] \overline{E}_i(El,Az)$$

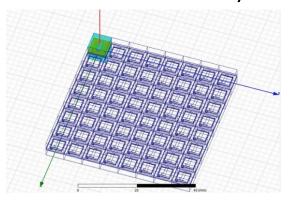




IC design



64 element antenna array

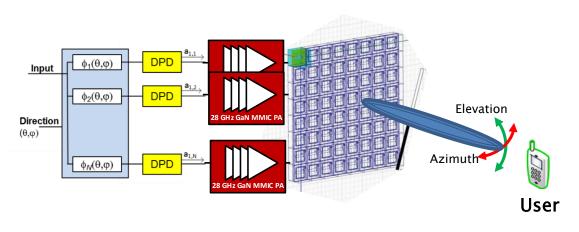


- PA model extracted from IC CAD
- Antenna parameters from EM CAD
- Each PA perfectly linearized for 50Ω load ($a_2 = 0$)
- Ideal phased array beam steering. No amplitude tapering

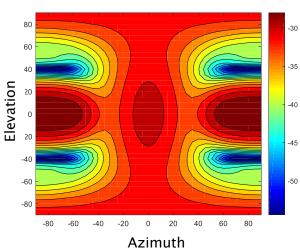
[C. Fager et al., "Analysis of Nonlinear Distortion in Phased Array Transmitters," Proc. INMMiC, 2017]







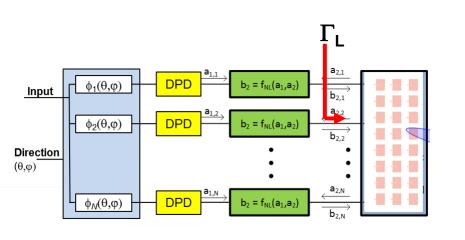
User EVM vs. scan direction



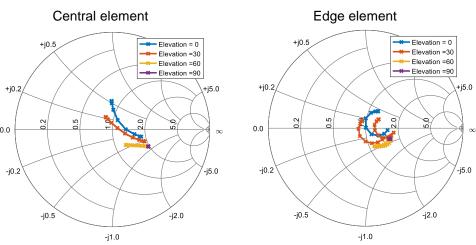
Distortion highly direction dependent. Why?







 Γ_L vs. scan direction

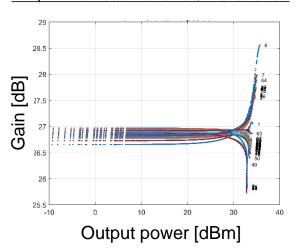


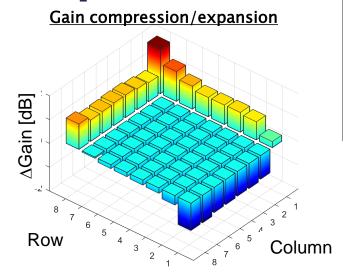
Significant variation of PA load impedance vs. beam steering

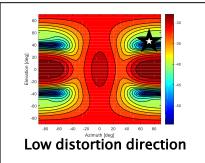




AM/AM for each of the 64 branches





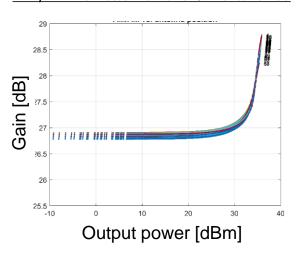


Distortion averaging effects happening inside the array

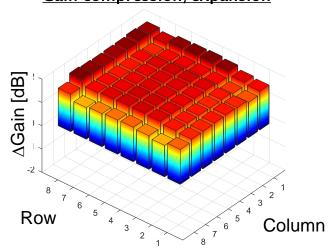


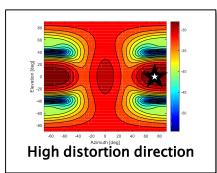


AM/AM for each of the 64 branches



Gain compression/expansion



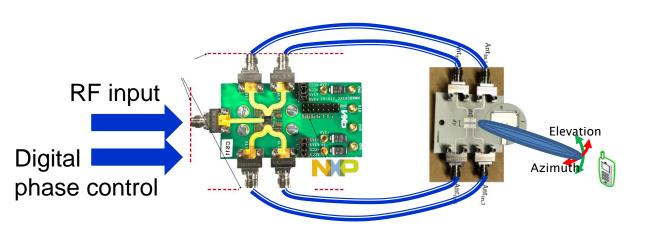


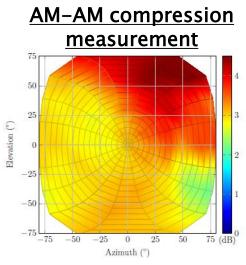
- Distortion addition for some directions
- Direction dependent user distortion ightarrow Direction dependent DPD needed





Experimental verification



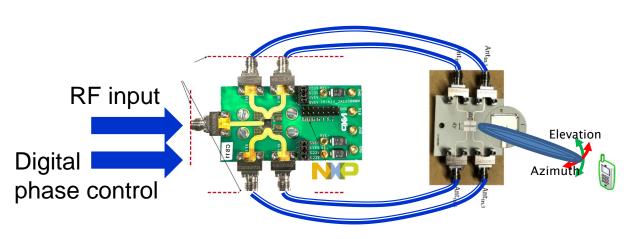


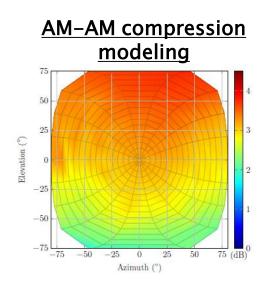
- Nonlinear effects observed also in mm-wave measurements
- 30 GHz beamforming chip





Experimental verification



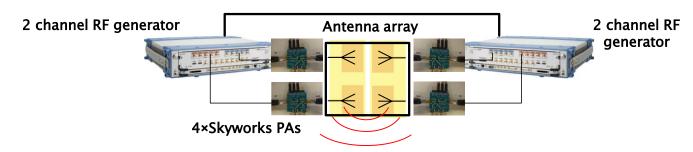


- Load-pull based nonlinear modeling
- 4 element patch antenna: EM simulation based modeling
- Reasonable modeling accuracy

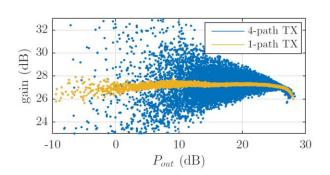




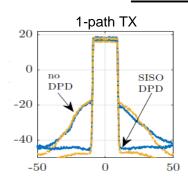
MIMO transmitter distortion

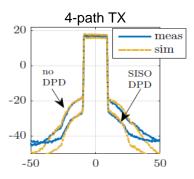


Nonlinear PA-antenna interactions



Far field distortion



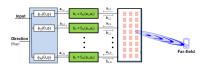


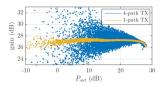
 $[K.\ Hausmair\ et\ al.\ "Prediction\ of\ Nonlinear\ Distortion\ in\ Wideband\ Active\ Antenna\ Arrays,"\ IEEE\ T-MTT,\ 2017]$

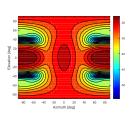




Summary – transmitter RF modeling

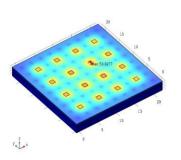






- Framework for efficient simulation of active antenna systems
- Improved understanding of circuitsantenna interactions with realistic signals
- New nonlinear effects predicted in phased array and MIMO systems

ELECTRO-THERMAL SIMULATIONS



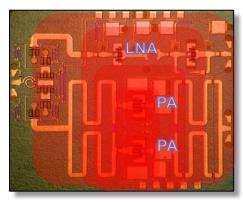
Thermal modeling
Power dissipation modeling
mm-wave transmitter example

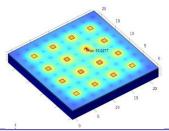




Heating concerns

- Heat concentration in active antenna arrays
- Chip level heating effects
 - Thermal coupling
 - Efficient power amplifiers
- System level effects
 - Performance degradation
 - Reliability



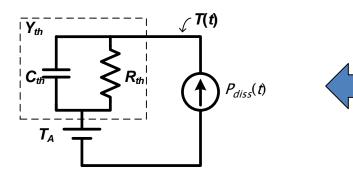


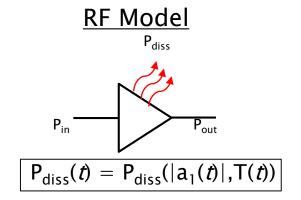




Linear heating model

Thermal model



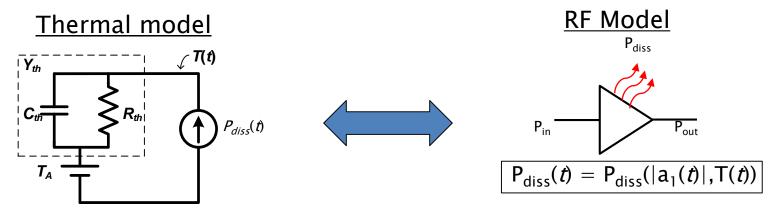


- $T(t) = T_A + P_{diss}(t) * z_{th}$
- Thermal admittance: $Y_{th} = G_{th} + j\omega C_{th}$





Linear heating model



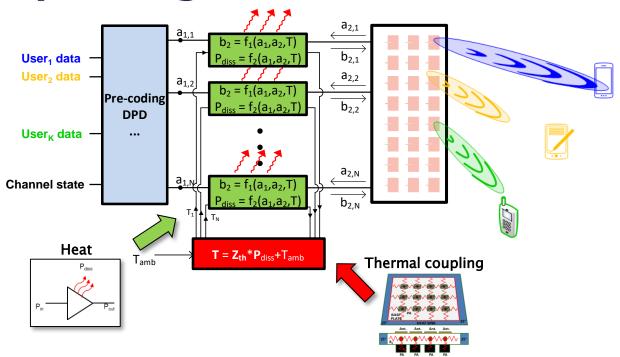
Envelope time-stepped solution for T(t)

$$\mathbf{T}_{n+1} = T_{\text{amb}} + \left(\mathbf{G}_{\text{th}} + 2\pi f_s \mathbf{C}_{\text{th}}\right)^{-1} \left(\mathbf{P}_{\text{diss},n} + 2\pi f_s \mathbf{C}_{\text{th}} \left(\mathbf{T}_n - T_{\text{amb}}\right)\right)$$





Incorporating thermal effects



[C. Fager et al. "Analysis of Thermal Effects in Active Antenna Array Transmitters...," Proc. INMMiC, 2015]



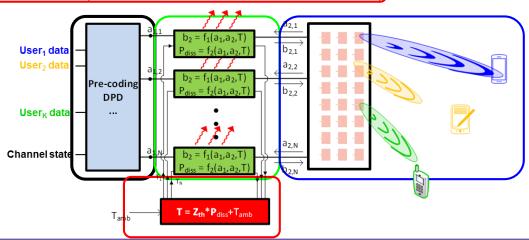
Combined RF/EM/Thermal simulation

$$\mathbf{a}_{1,n} = \mathbf{G}\mathbf{x}_n$$

$$\mathbf{b}_{2,n} = \mathbf{S}_{21}(|\mathbf{a}_{1,n}|, \mathbf{T}_n)\mathbf{a}_{1,n} + \mathbf{S}_{22}(|\mathbf{a}_{1,n}|, \mathbf{T}_n)\mathbf{S}_{ant}\mathbf{b}_{2,n} + \mathbf{T}_{22}(\mathbf{a}_{1,n}, \mathbf{T}_n)\mathbf{S}_{ant}^*\mathbf{b}_{2,n}^* + \mathbf{\rho}_{n-1}$$

$$E_n(\theta,\varphi) = \mathbf{b}_{2,n}^T \mathbf{\bar{E}}(\theta,\varphi)$$

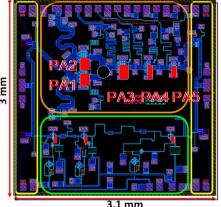
$$\mathbf{T}_{n+1} = T_{\text{amb}} + \left(\mathbf{G}_{\text{th}} + 2\pi f_s \mathbf{C}_{\text{th}}\right)^{-1} \left(\mathbf{P}_{\text{diss},n} \left(\left|\mathbf{a}_{1,n}\right|, \mathbf{T}_n\right) + 2\pi f_s \mathbf{C}_{\text{th}} \left(\mathbf{T}_n - T_{\text{amb}}\right)\right)$$



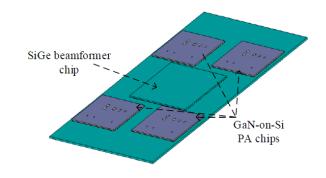


Example: SERENA 4-channel transmitter





- 1. PA1: $8 \times 115 \, \mu \text{m} = 920 \, \mu \text{m} \, (\approx 0.255 w_{\text{tot}})$
- 2. PA2: $8 \times 115 \, \mu m = 920 \, \mu m \, (\approx 0.255 w_{tot})$
- 3. PA3: $8 \times 115 \, \mu \text{m} = 920 \, \mu \text{m} \, (\approx 0.255 w_{\text{tot}})$
- 4. PA4: $8 \times 57 \, \mu \text{m} = 456 \, \mu \text{m} \, (\approx 0.126 w_{\text{tot}})$
- 5. PA5: $6 \times 65 \, \mu \text{m} = 390 \, \mu \text{m} \, (\approx 0.108 w_{\text{tot}})$

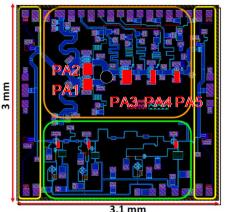


Material	Density $ ho ({ m kg/m^3})$	Thermal conductivity $k (W/m \cdot K)$	Specific heat $C_p\left(J/kg\!\cdot\!K\right)$
Megtron 7	1820	0.4	0.88
Si	2330	148	0.7
SiGe	3950	8.8	0.5

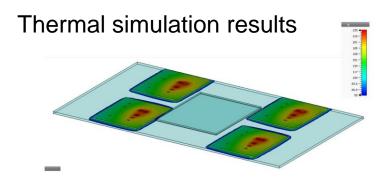


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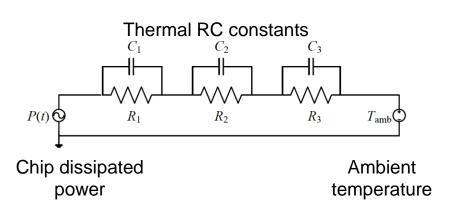
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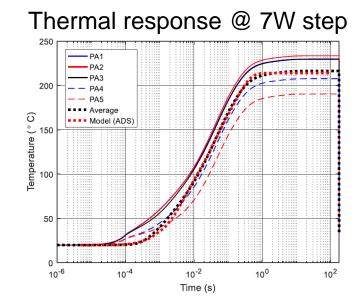




Thermal RC modeling

One GaN chip in a simplified package environment

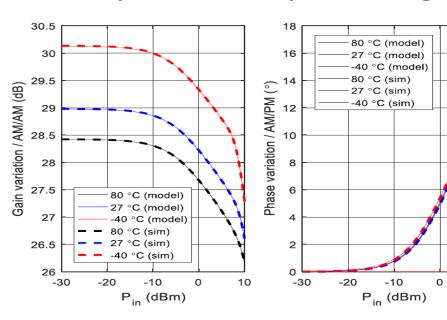






Power amplifier modeling

Temperature dependent gain



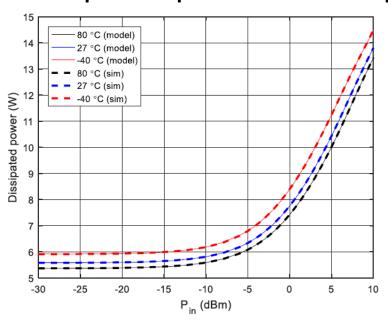
$$b_2(a_1, T) = \sum_{p_1=1}^{P_1} \alpha_{p_1}(T) a_1 |a_1|^{2(p_1-1)}$$

10



Power amplifier modeling

• Dissipated power vs. temperature



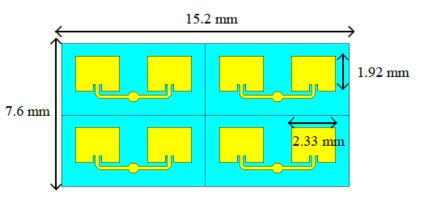
$$P_{\text{diss}}(|a_1|, T) = P_{\text{dc}} + P_{\text{in}} - P_{\text{out}}$$

= $\sum_{p_d=0}^{P_d} \xi_{p_d}(T)|a_1|^{p_d}$



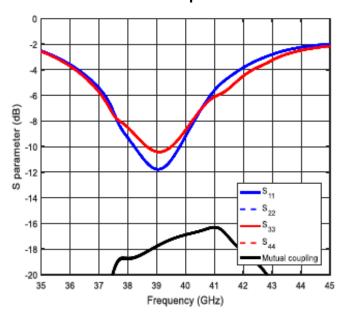


Antenna modeling



 Mismatch and mutualcoupling neglected in model

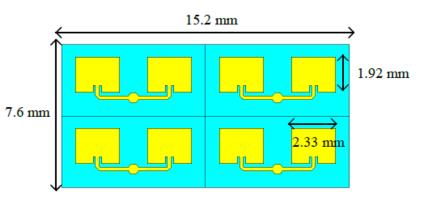
Simulated S-parameters





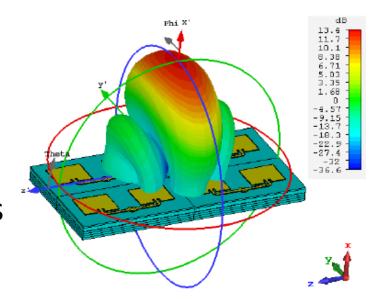


Antenna modeling



 Embedded element patterns for unity excitations

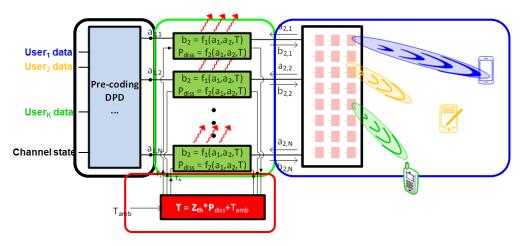
Far-field radiation patterns



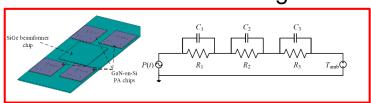




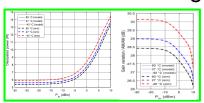
Electro-thermal simulation framework



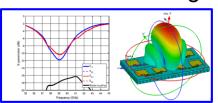
Thermal modeling



RF circuit modeling



Antenna modeling

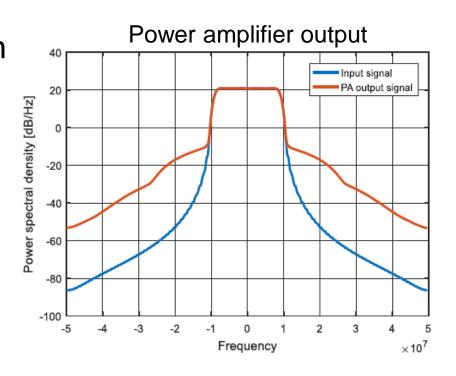






Prediction of transmitter RF nonlinearities

- PA input-/output spectrum for modulated signals
- PA-to-PA nonlinear interactions

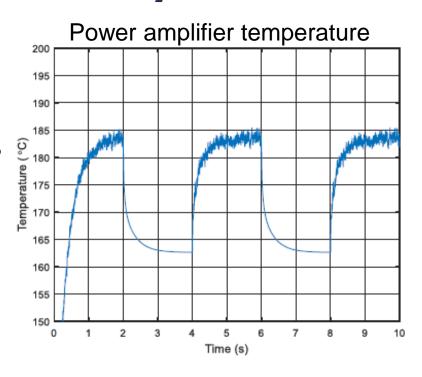






Prediction of PA temperature dynamics

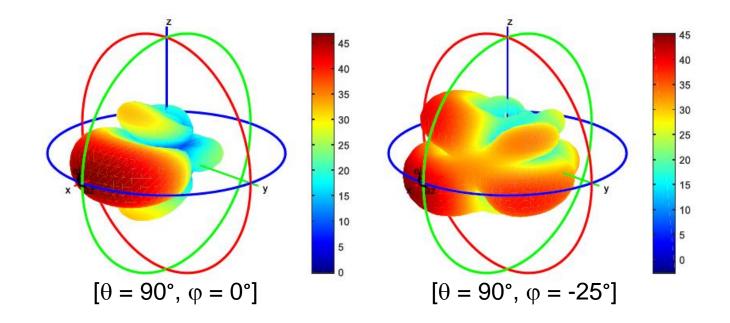
- Thermal transients
- On-off switching, e.g.
 between T/R in TDD systems
- Heat spreading in arrays







Prediction of radiation patterns

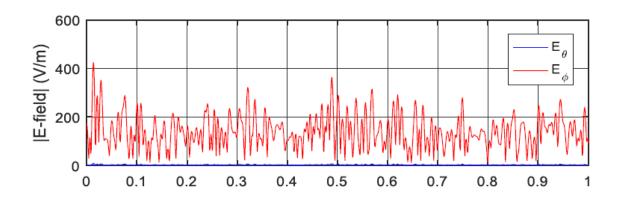






Prediction of modulated field at user

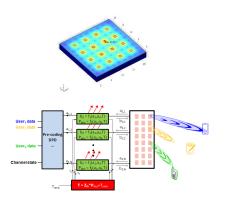
 Modulated fields in both polarizations considering both PA, thermal, antenna and beamforming settings

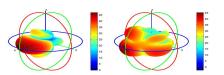






Summary – Electro-thermal simulations





- Thermal effects at circuit- and package levels
- Framework for electro-thermal transmitter simulations
- Prediction of joint electrical- and thermal effects in mm-wave antenna arrays

CONCLUSIONS





Conclusions

- mm-wave transmitter design is cross-disciplinary
 - Co-design between signals, circuits & antennas
- New nonlinear distortion phenomena
 - Both circuit and array level linearization needed
 - Low complexity MIMO linearization proposed
- Power dissipation a great challenge
 - Thermal coupling effects
- Understanding linearity and power dissipation effects through accurate multi-physics simulations is critical for successful 5G system design

<u>Acknowledgments</u>















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